

# From Volterra series through block-oriented approach to Volterra series

**Pawel Wachel and Przemyslaw Sliwinski**

Department of Control Systems and Mechatronics  
Wroclaw University of Technology

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# Outline

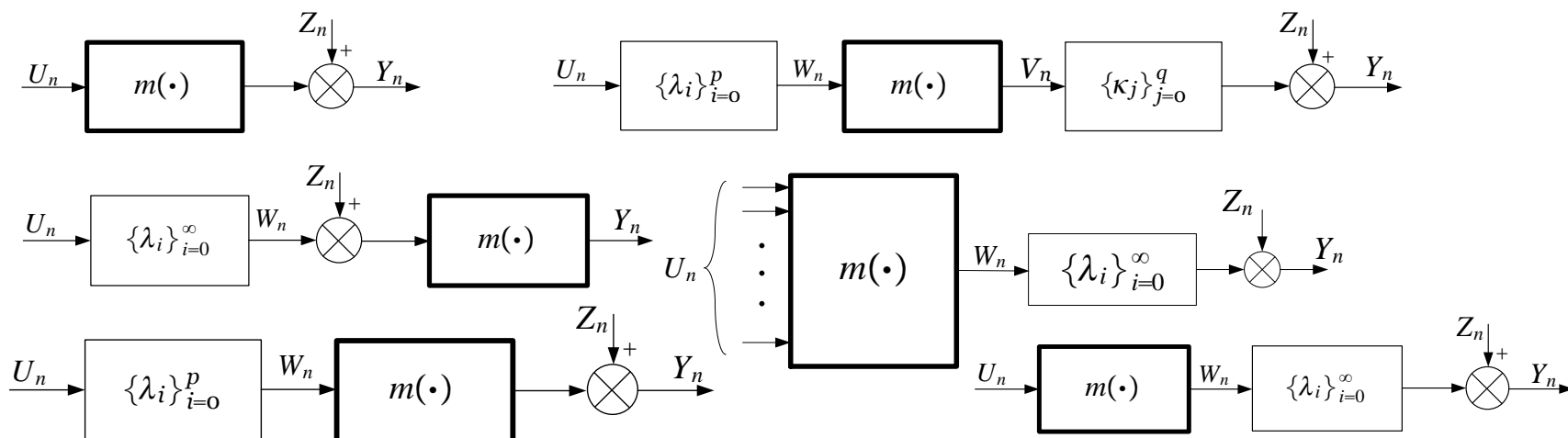
- Introduction – Volterra series and block oriented models
- Nonparametric kernel methods – basic concepts
- Nonparametric kernel regression in application to system identification
- Mixed (parametric–nonparametric) system identification
- Aggregative modeling – dictionary approach
- Application of aggregative approach to Volterra series modeling of LNL system

# Volterra vs. block-oriented approach

## Classic Volterra approach



## Block-oriented „alternatives”



# General block-oriented identification setup

**Input signal**

*i.i.d. random sequence*

**Nonlinearity**

*Nonparametric prior knowledge*

**Dynamic subsystem(s)**

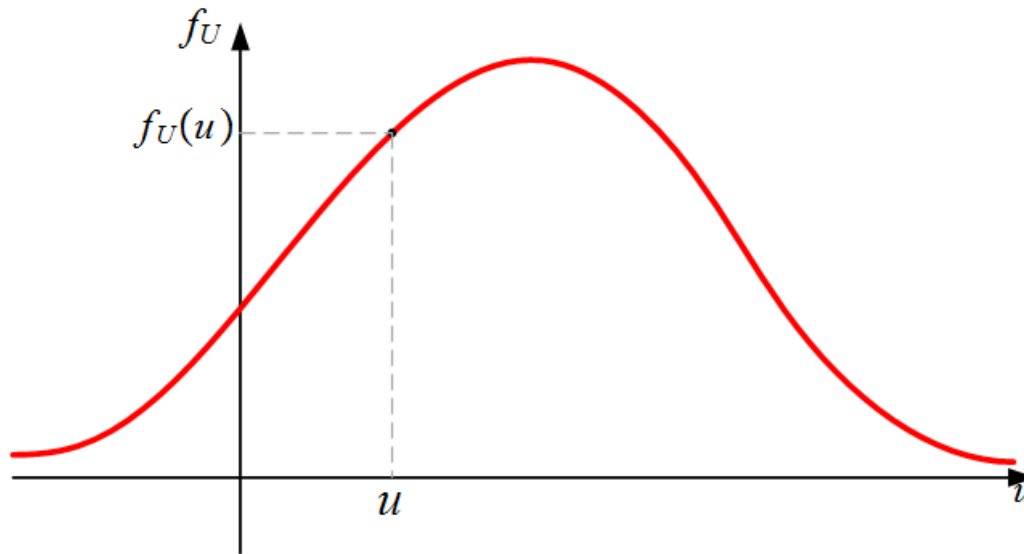
*Asymptotic stability assumed*

# Nonparametric kernel density estimation

- Let us consider the following estimation problem: Given the set of *i.i.d.* measurements

$$U_1, U_2, \dots, U_N$$

with the **unknown** probability density function  $f_U(\cdot)$

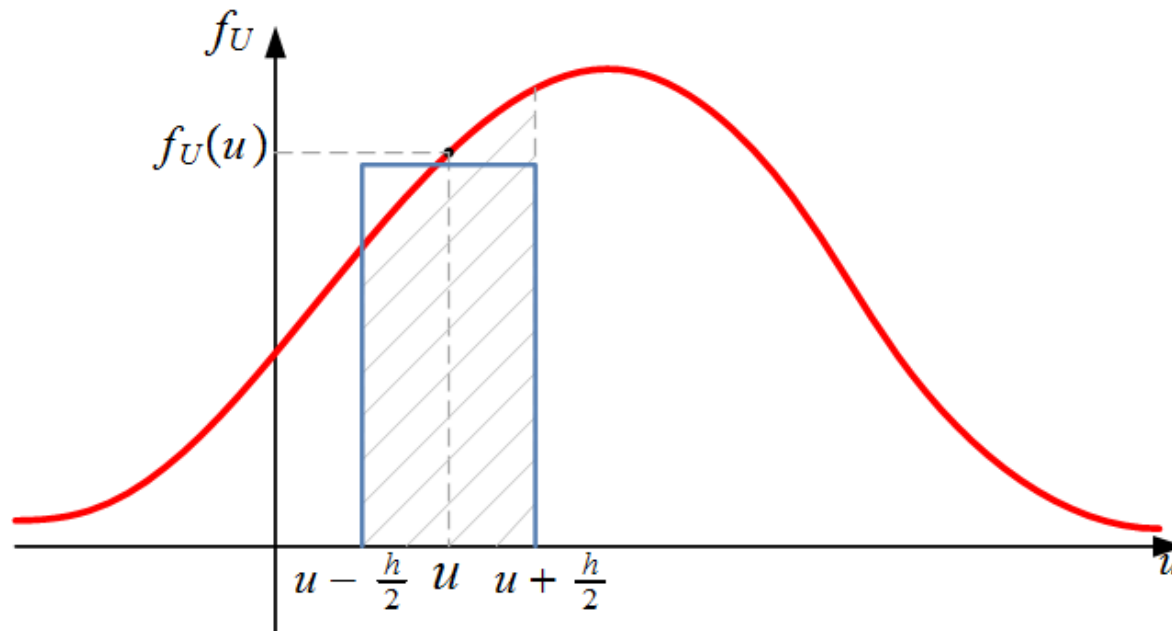


find (estimate)  $f_U(\cdot)$  in some point  $u$  from the set  $U_1, U_2, \dots, U_N$

# Nonparametric kernel density estimation

- The main idea behind the considered estimate is based on the observation that for any  $u$  and relatively small constant  $h$ :

$$f_U(u) \approx \frac{1}{h} \int_{u-h/2}^{u+h/2} f_U(v) dv.$$

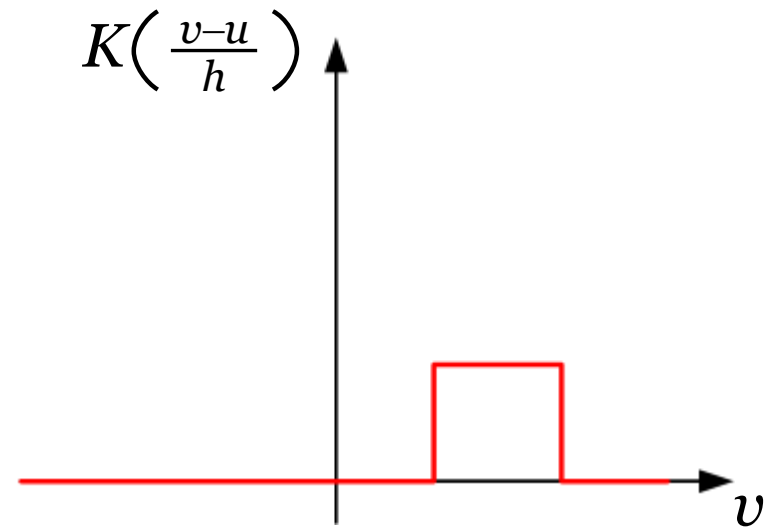
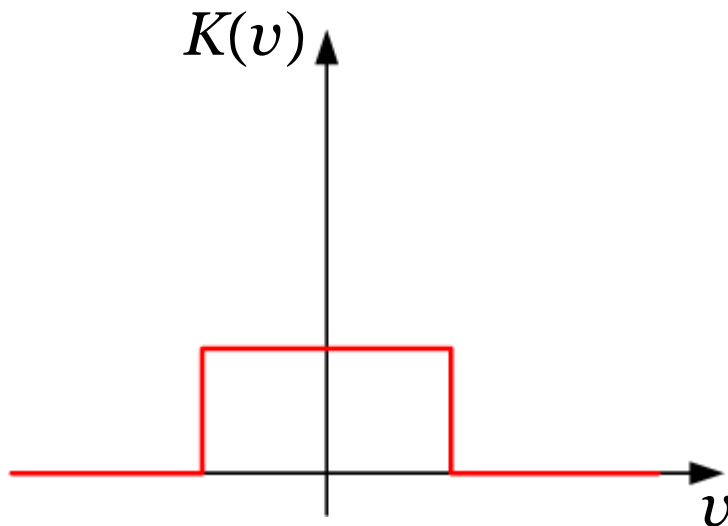


# Nonparametric kernel density estimation

- Let now  $K(\cdot)$  be a box kernel function of the form

$$K(v) = \begin{cases} 1 & \text{for } v \in \left\langle -\frac{1}{2}; \frac{1}{2} \right\rangle \\ 0 & \text{for } v \notin \left\langle -\frac{1}{2}; \frac{1}{2} \right\rangle \end{cases}$$

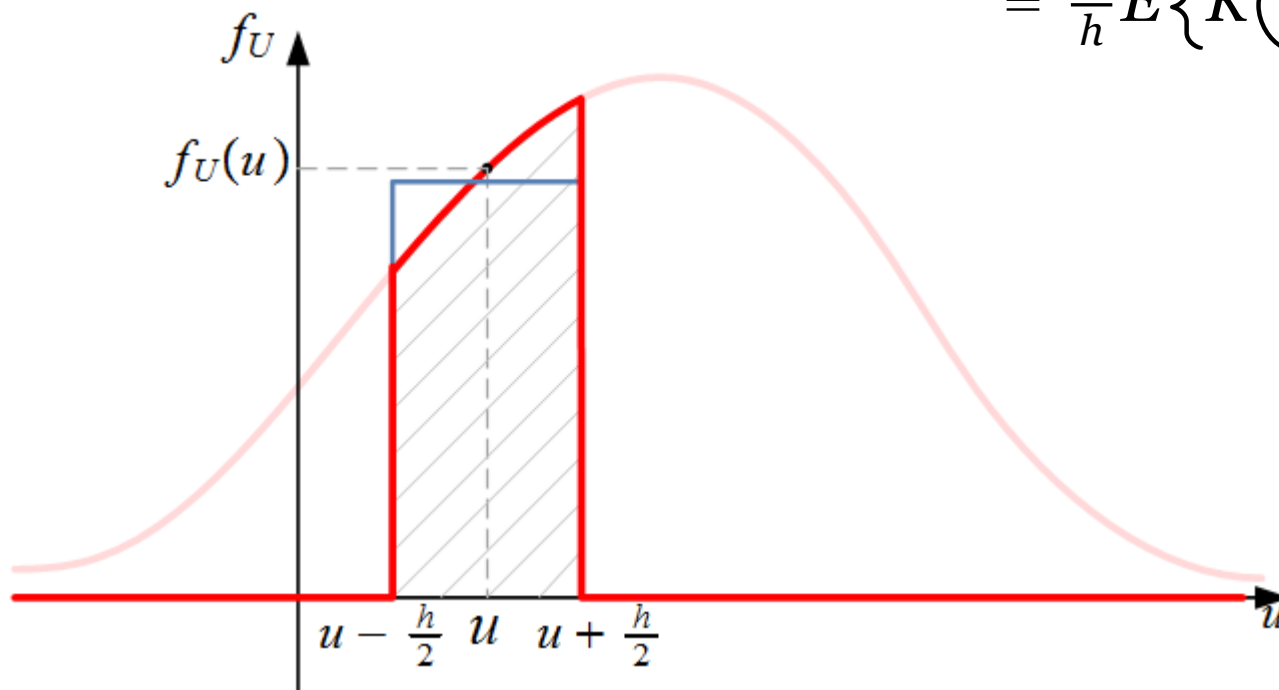
**How the kernel works?**



# Nonparametric density estimation

- Using the kernel function we obtain further

$$f_U(u) \approx \frac{1}{h} \int_{u-h/2}^{u+h/2} f_U(v) dv = \frac{1}{h} \int_{-\infty}^{+\infty} f_U(v) K\left(\frac{v-u}{h}\right) dv$$
$$= \frac{1}{h} E\left\{K\left(\frac{U_1-u}{h}\right)\right\}$$





# Nonparametric density estimation

- Finally, on the basis of

$$f_U(u) \approx \frac{1}{h} E \left\{ K \left( \frac{U_1 - u}{h} \right) \right\}$$

we obtain the following Kernel Density Estimate (Rosenblatt, Parzen)

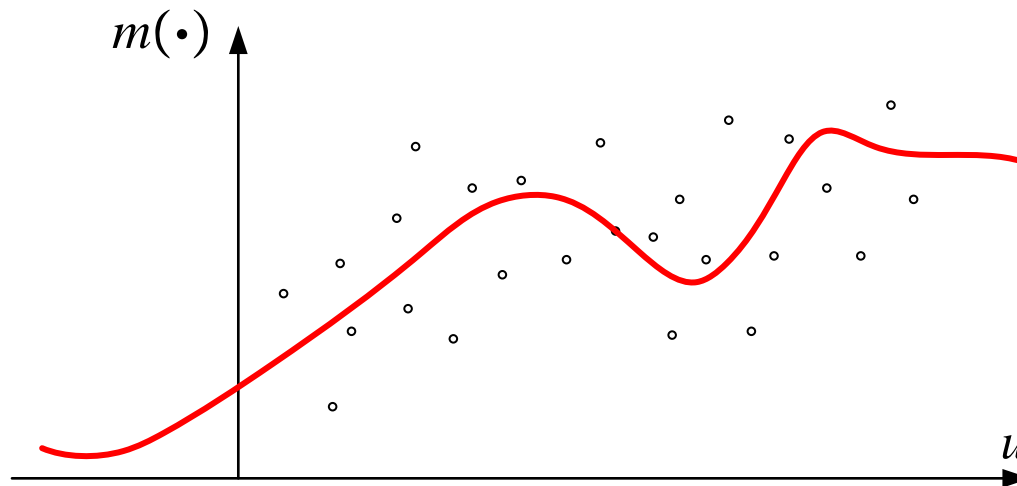
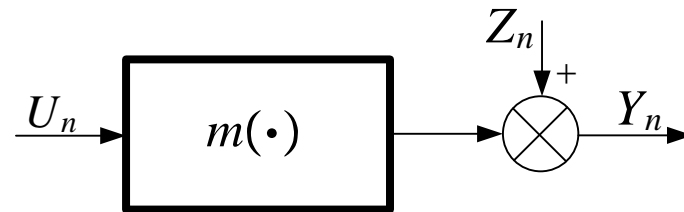
$$\hat{f}_U(u) = \frac{1}{Nh} \sum_{i=1}^N K \left( \frac{U_i - u}{h} \right)$$

(which is a sum of scaled and translated kernel functions)

# Nonparametric kernel identification

- Consider a static nonlinear system described by the formula

$$Y_n = m(U_n) + Z_n$$

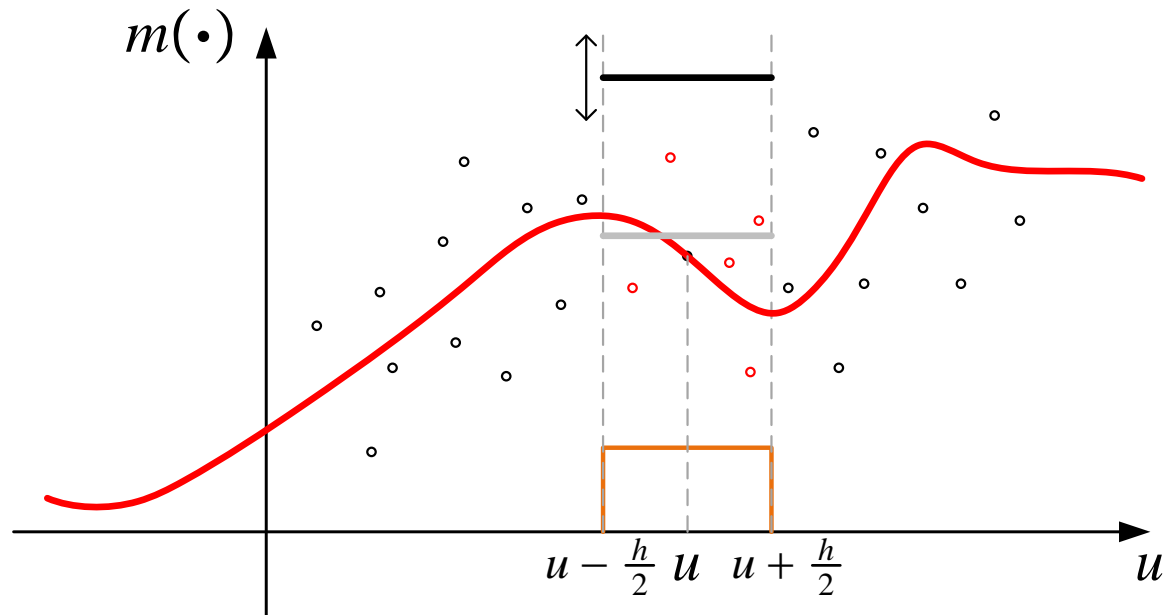


# Nonparametric kernel identification

- Consider the following local least squares quality function

$$\tilde{Q}(c; u) = \sum_{i=1}^N [Y_i - c]^2 K\left(\frac{U_i - u}{h}\right)$$

- Goal: find  $c_0$  such that  $c_0 = \operatorname{argmin}_{c \in \mathbb{R}} \tilde{Q}(c; u)$

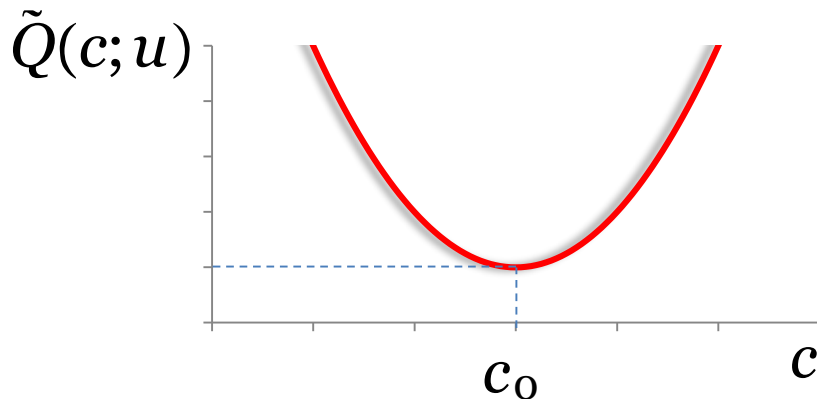


# Nonparametric kernel identification

$$\tilde{Q}(c; u) = \sum_{i=1}^N [Y_i - c]^2 K\left(\frac{U_i - u}{h}\right)$$



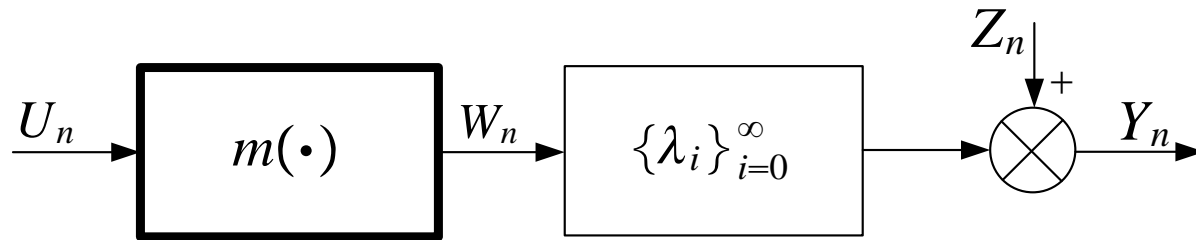
$\tilde{Q}(c; u)$  is a (non-negative) **quadratic** function of variable  $c$ .



**Solution:** Nadaraya–Watson kernel estimate

$$c_0 = \hat{m}(u) = \frac{\sum_{i=1}^N Y_i K\left(\frac{U_i - u}{h}\right)}{\sum_{i=1}^N K\left(\frac{U_i - u}{h}\right)}$$

# Hammerstein system – nonlinearity estimation



$$W_n = m(U_n),$$

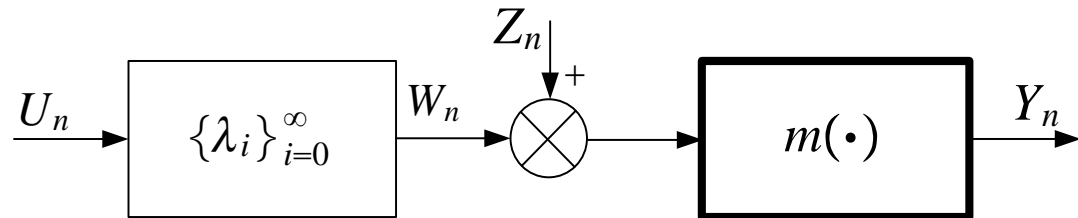
$$Y_n = \sum_{i=0}^{\infty} \lambda_i W_{n-i} + Z_n$$

$$\hat{\mu}(u) = \frac{\sum_{i=1}^N Y_i K\left(\frac{U_i - u}{h}\right)}{\sum_{i=1}^N K\left(\frac{U_i - u}{h}\right)}$$

$\hat{\mu}(u) \rightarrow \alpha \cdot m(u) + \beta$ , as  $N \rightarrow \infty$  in probability

# Wiener system

Gaussian input assumed



$$W_n = \sum_{l=0}^{\infty} \lambda_l U_{n-l} + Z_n,$$

$$Y_n = m(W_n).$$

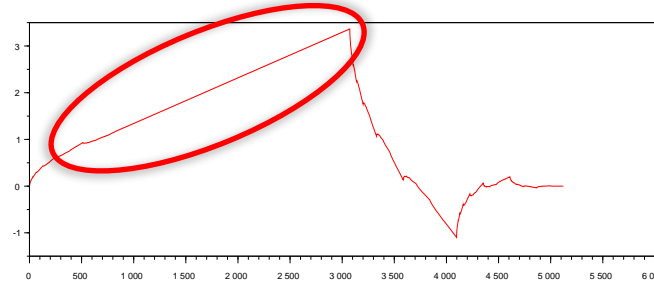
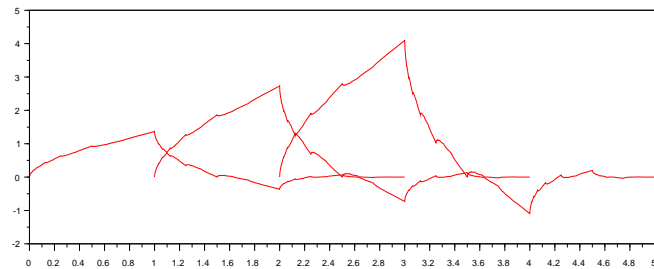
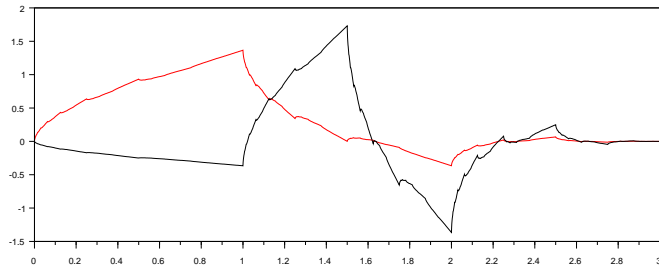
**Main concept:** Replace input measurements with output ones in the kernel estimate.

$\hat{\mu}(y) \rightarrow \alpha m^{-1}(y)$  as  $N \rightarrow \infty$  in probability.

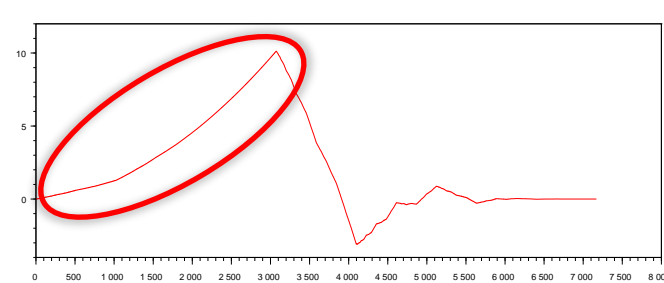
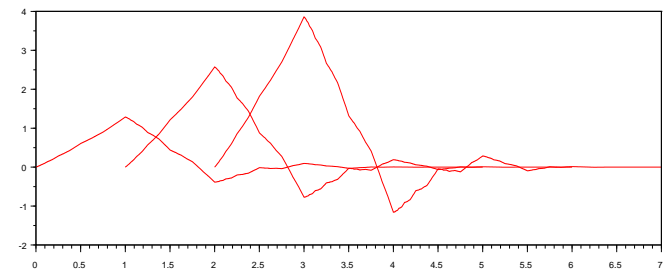
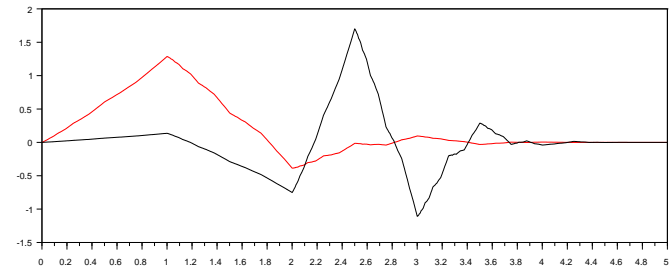
$$\hat{\mu}(y) = \frac{\sum_{i=1}^N U_i K\left(\frac{Y_i - y}{h}\right)}{\sum_{i=1}^N K\left(\frac{Y_i - y}{h}\right)}$$

# Orthogonal algorithms and wavelets

## 2nd order Daubechies



## 3rd order Daubechies

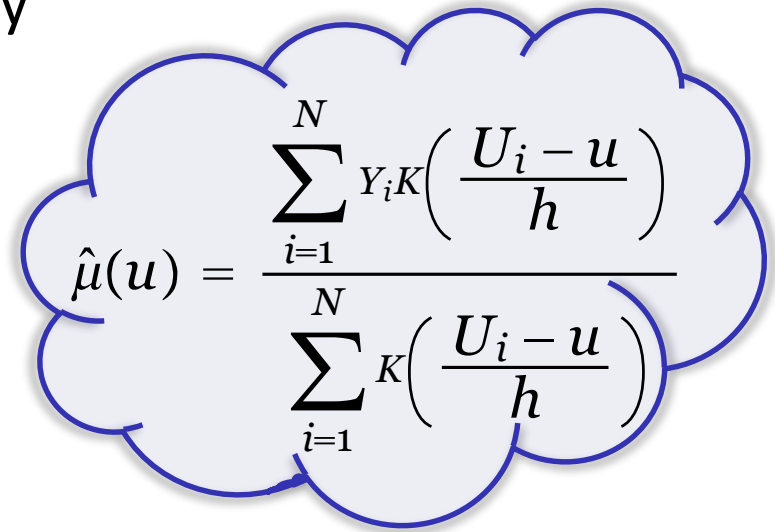


# Orthogonal wavelet regression estimate

Orthogonal estimate of the nonlinearity

$$\hat{\mu}_K(u) = \frac{\sum_{i=1}^N Y_i \mathfrak{g}_K(u, U_i)}{\sum_{i=1}^N \mathfrak{g}_K(u, U_i)}$$




$$\hat{\mu}(u) = \frac{\sum_{i=1}^N Y_i K\left(\frac{U_i - u}{h}\right)}{\sum_{i=1}^N K\left(\frac{U_i - u}{h}\right)}$$

where

$$\mathfrak{g}_K(u, v) = \sum_{n=n_1}^{n_2} \varphi_{Kn}(u) \varphi_{Kn}(v)$$

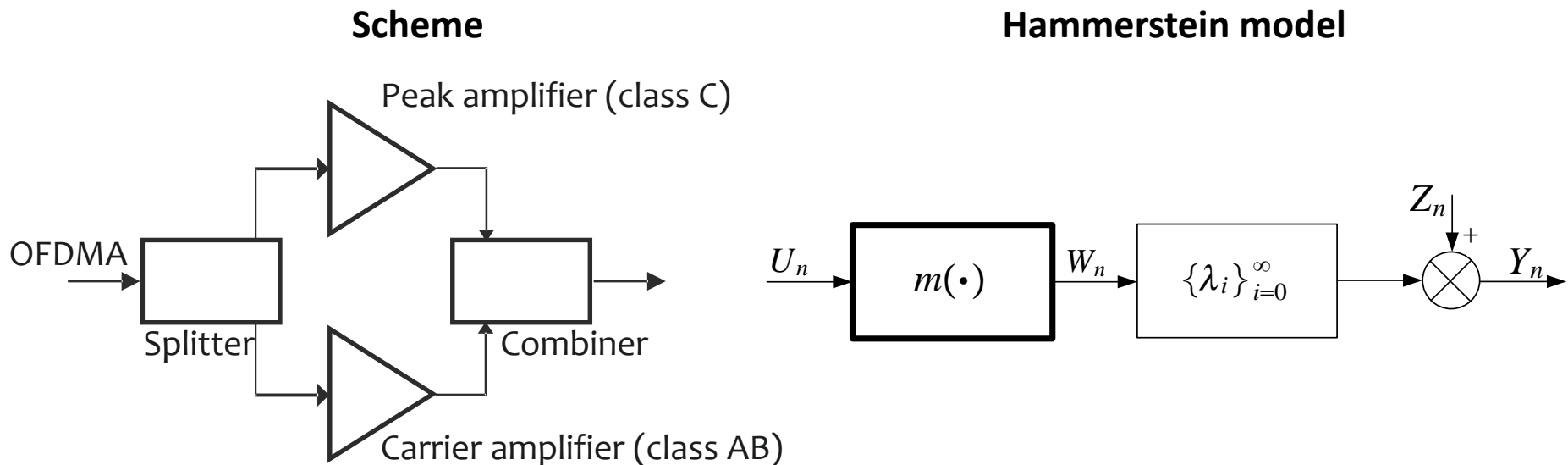
and  $\{\varphi_{Kn}(\cdot)\}$  is a family of orthogonal functions.



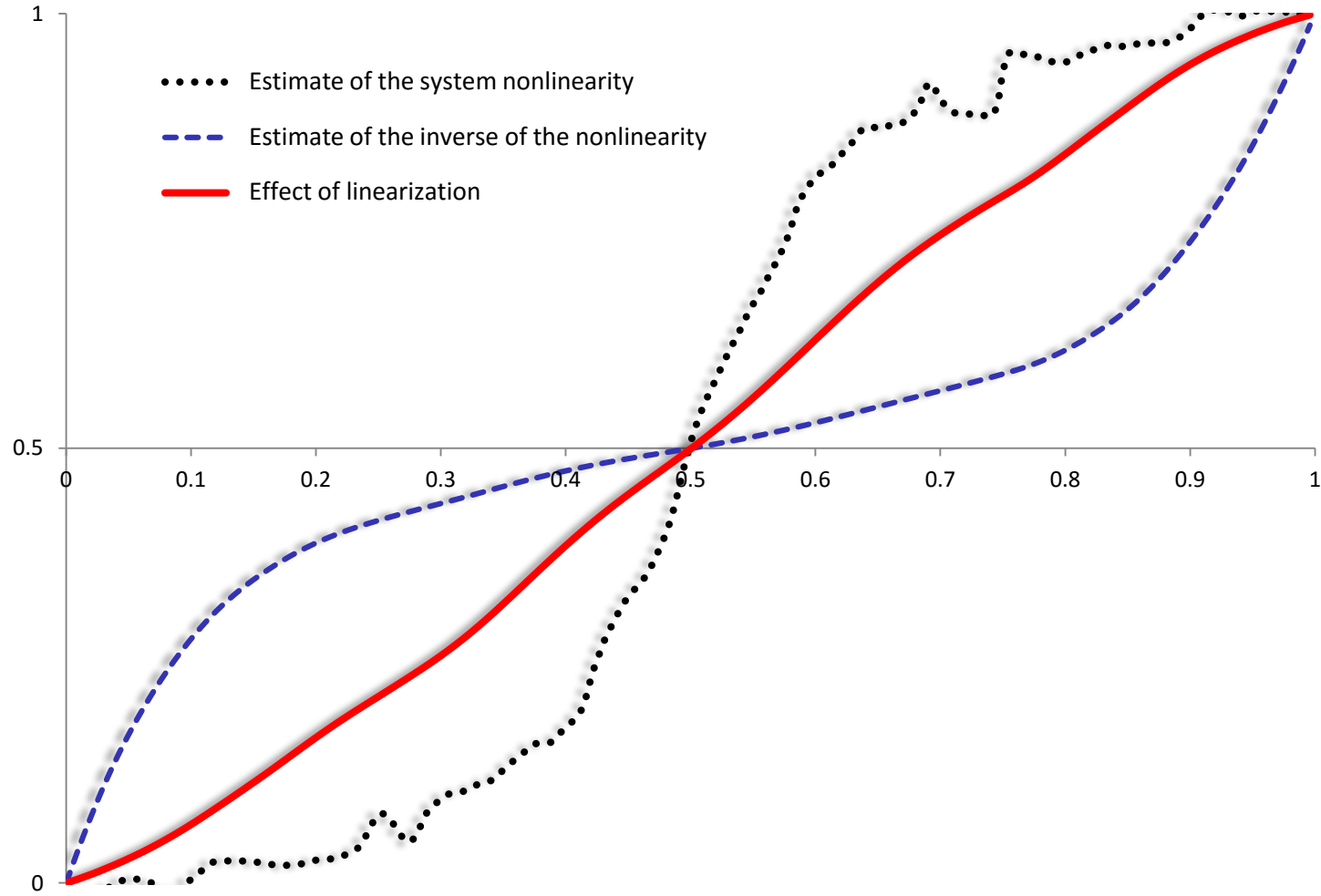
# Application to predistortion – Doherty amplifier

**Predistortion problem:** Estimate an **inverse** of the system's nonlinearity  $m^{-1}(\cdot)$  and linearize system by replacing input  $U_n$  with  $m^{-1}(U_n)$ .

## Doherty amplifier



# Predistortion – practical example

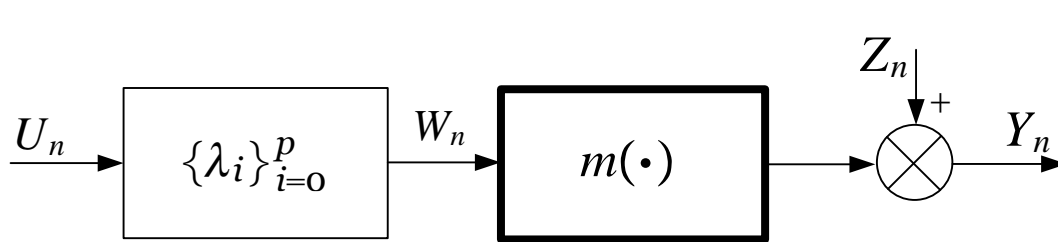


# Average Derivative Estimation – Wiener system

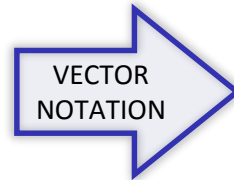
**Fact (chain rule):** Let  $m(\cdot)$  be a differentiable function, then

$$\frac{d}{dx} m(\alpha x) = \alpha \left[ \frac{d}{dx} m(\alpha x) \right]$$

**Conclusion:** Derivative can **extract** factor  $\alpha$  embedded in the argument of  $m(\cdot)$ .



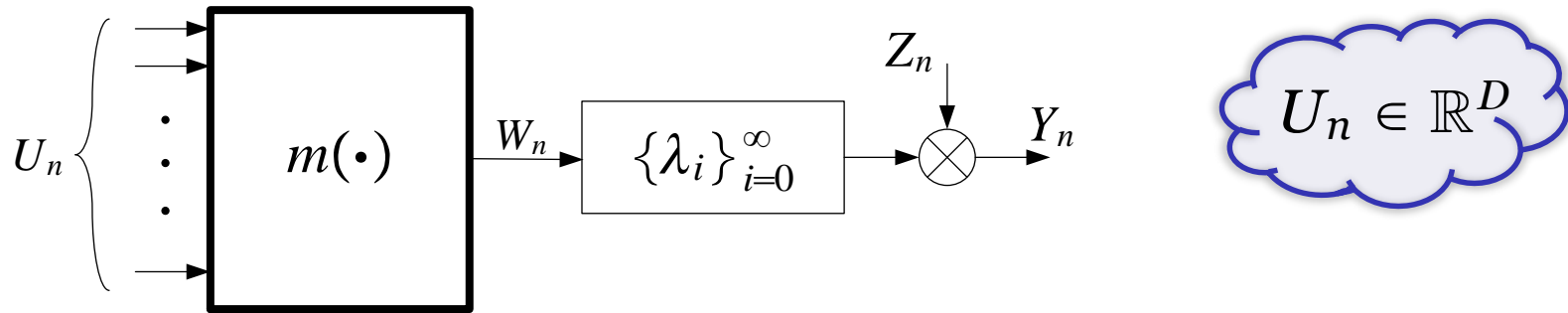
$$Y_n = m\left(\sum_{i=0}^p \lambda_i U_{n-i}\right) + Z_n$$



$$Y_n = m(\boldsymbol{\lambda}^T \mathbf{U}_n) + Z_n$$

# Identification from structured input data

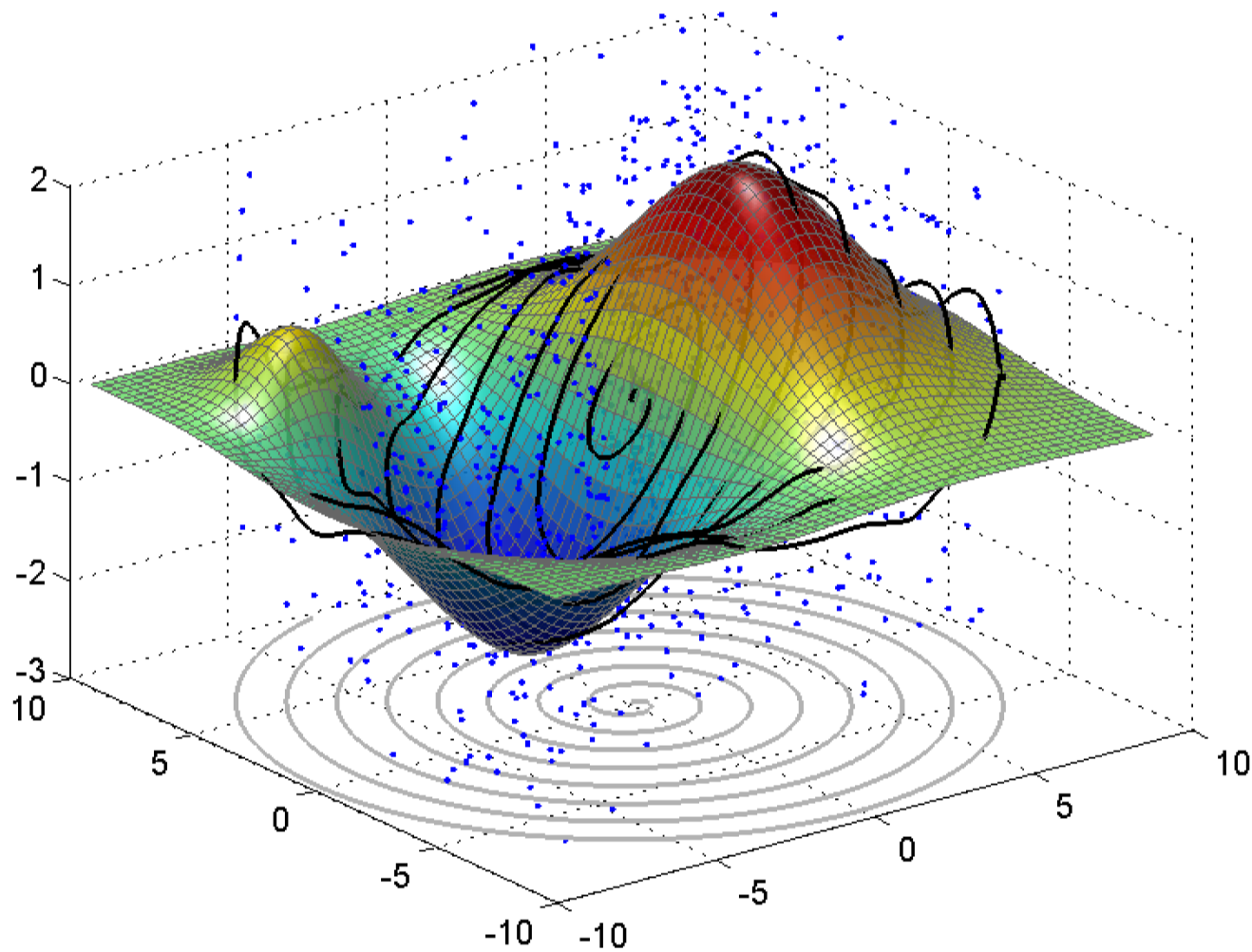
- Consider the following MISO Hammerstein system:



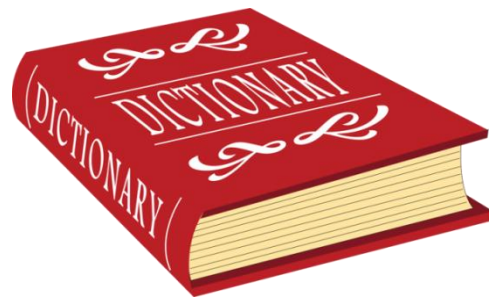
**Fact:** Convergence rate of the nonparametric estimators strongly depends on the number of system inputs.

**Assumption:** Input data are structured *i.e.* grouped on unknown  $d$ -dimensional manifold ( $d < D$ ).

# Identification from structured input data

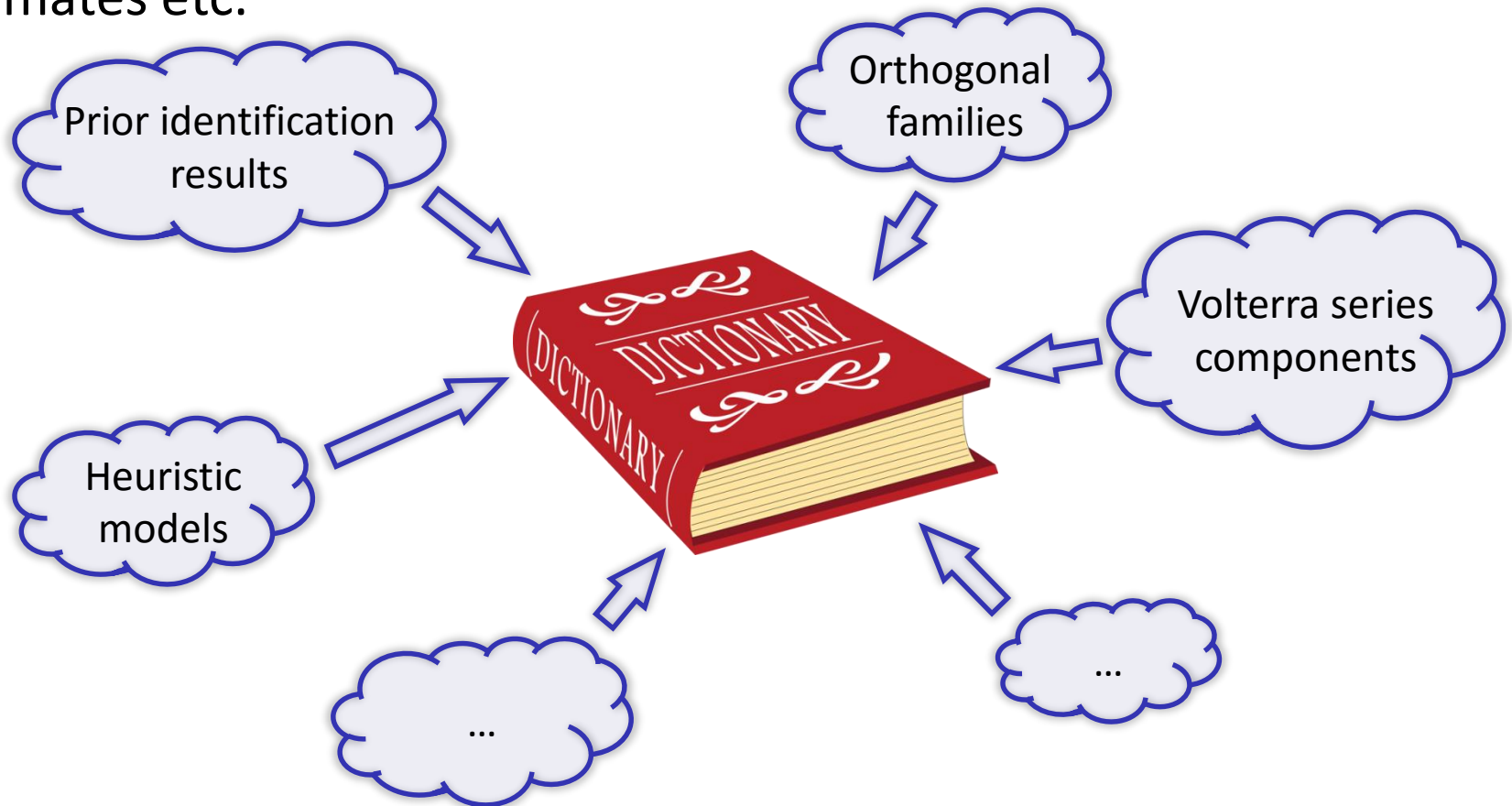


# AGGREGATIVE MODELING



# Dictionary modeling

The *a priori* knowledge about the system is **collected in a dictionary** in a form of the system components, models, pre-estimates etc.



# Aggregative modeling

- The class of considered SISO systems can be described by the general discrete-time input-output equation

$$Y_n = m(X_n, X_{n-1}, X_{n-2}, \dots, X_{n-p}) + Z_n,$$

## Assumptions:

- $\{X_n\}$  is a stationary sequence of independent random variables.
- $m(\mathbf{x})$  is **any** bounded function, such that  $|m(\mathbf{x})| \leq L < \infty$
- $\{Z_n\}$  is a zero-mean i.i.d. random sequence with finite variance.
- There is a dictionary of maps,

$$S = \{\bar{m}_i : \mathbb{R}^{p+1} \rightarrow [-L, L], i = 1, \dots, D\}, D > 2$$

collected by a user to model the system.



# Aggregative modeling

The algorithm combines all the dictionary entries into a single one, referred to as the aggregated empirical model.

$$m(\mathbf{x}; \hat{\boldsymbol{\alpha}}) = \sum_{i=1}^D \hat{\alpha}_i \bar{m}_i(\mathbf{x})$$



where

$$\hat{\boldsymbol{\alpha}} = \operatorname{argmin}_{\boldsymbol{\alpha}} \hat{Q}(\boldsymbol{\alpha}), \text{ subject to } \|\boldsymbol{\alpha}\|_1 \leq 1,$$

and where

$$\hat{Q}(\boldsymbol{\alpha}) = \frac{1}{N-p} \sum_{i=p+1}^N [m(\mathbf{X}_i; \boldsymbol{\alpha}) - Y_i]^2.$$

# Convergence

For a wide class of nonlinear systems and aggregative (dictionary) model, it holds that

$$E\{Q(\hat{\alpha})\} - Q(\alpha^*) \leq C \frac{\sqrt{N \ln D}}{N - p}$$

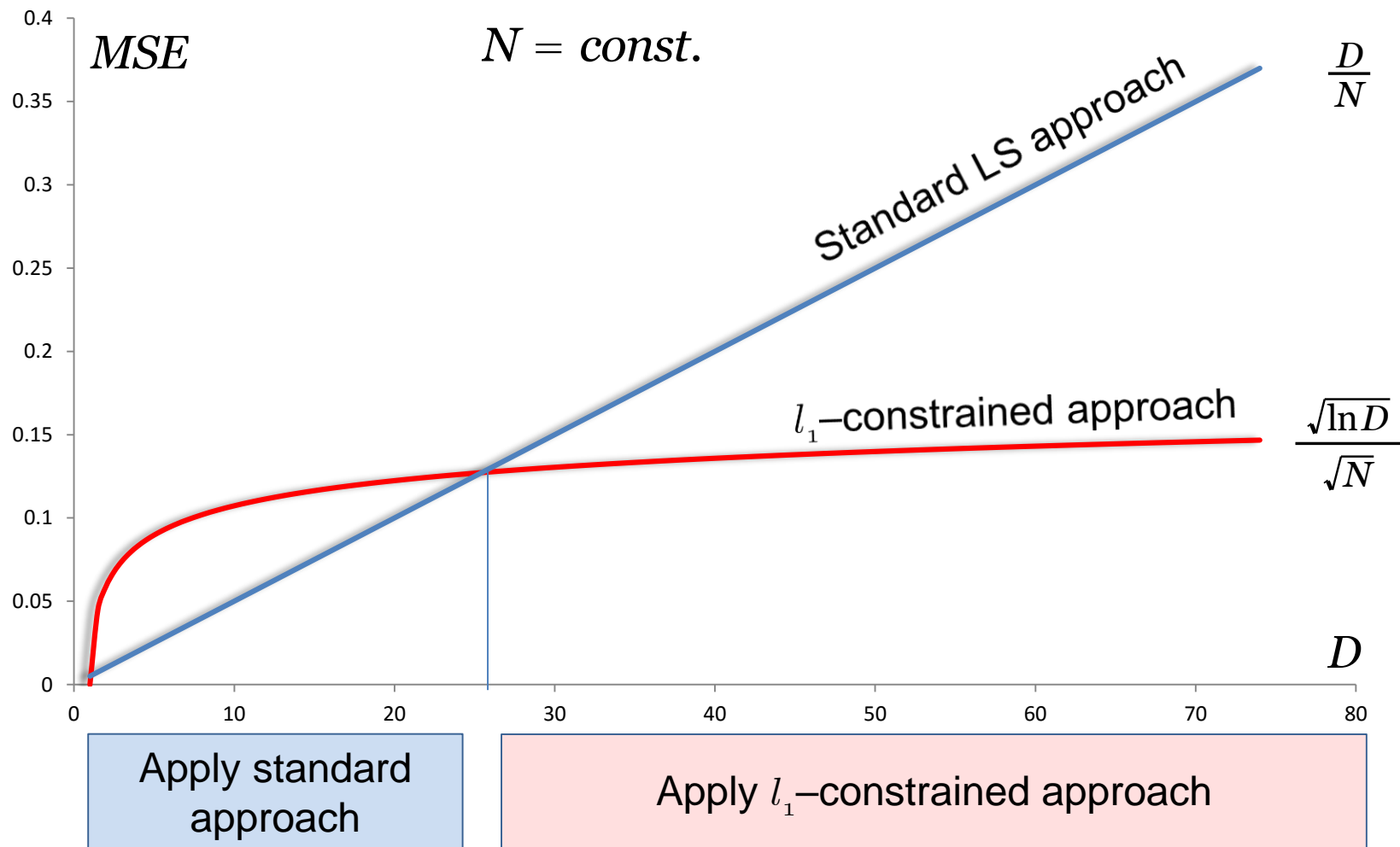
where

$$Q(\alpha) = E\{m(\mathbf{X}_n; \alpha) - Y_n\}^2$$

and where

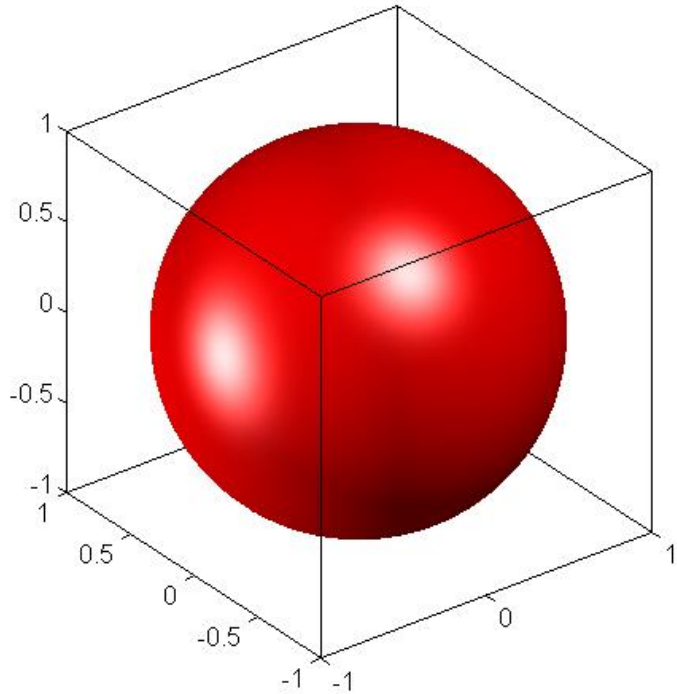
$$\alpha^* = \operatorname{argmin}_{\alpha \in A} Q(\alpha)$$

# $l_1$ -constrained vs. standard LS modeling...



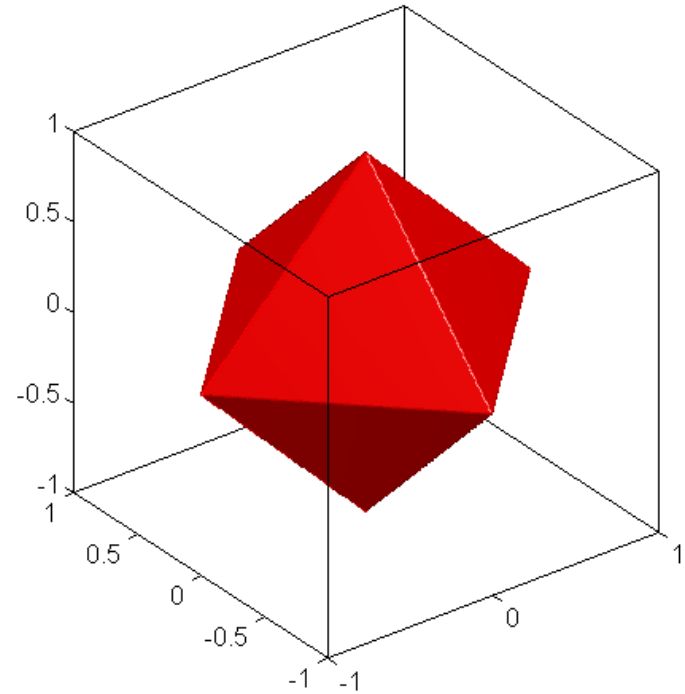
# Why it works?

$l_2$ -ball for  $D=3$



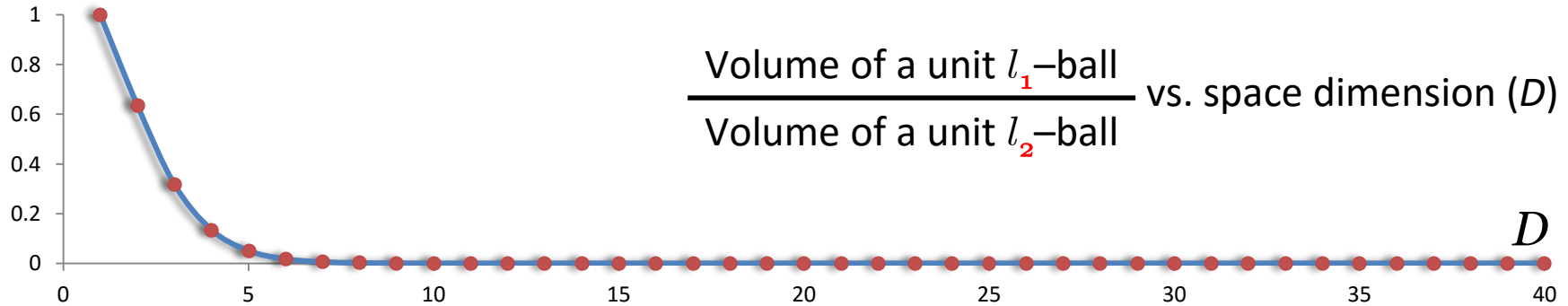
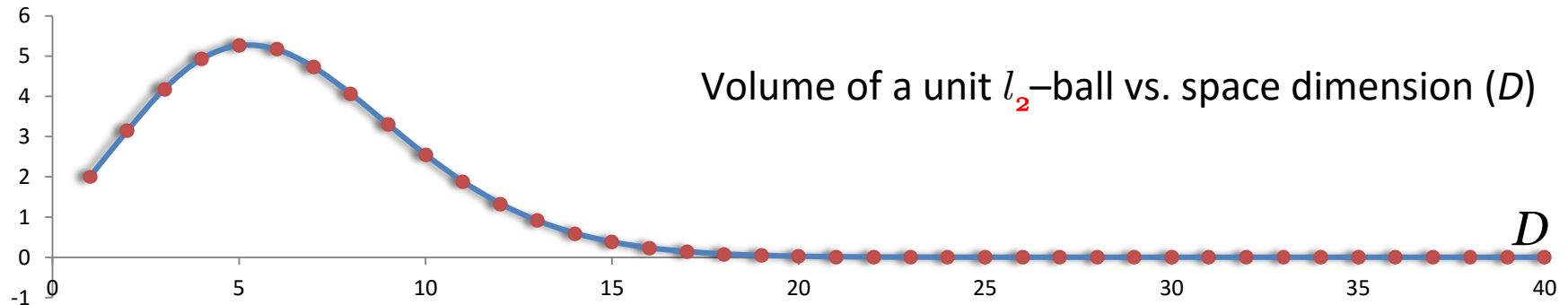
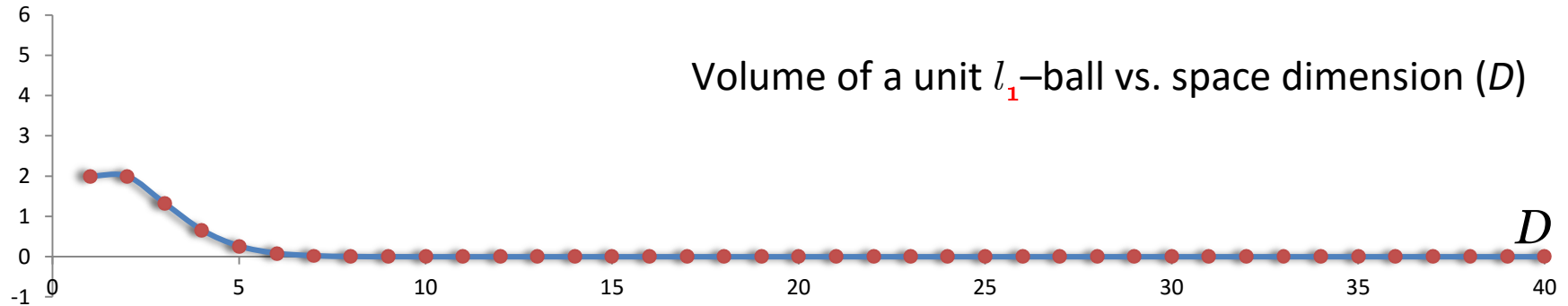
$$V = \frac{4}{3}\pi \approx 4.1888$$

$l_1$ -ball for  $D=3$



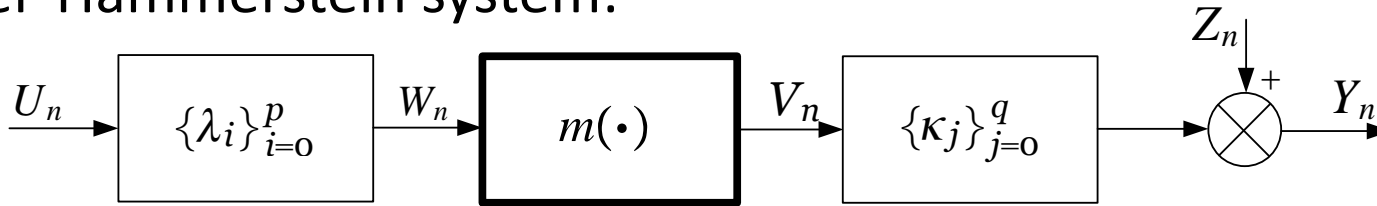
$$V = 1.333$$

# Why it works?



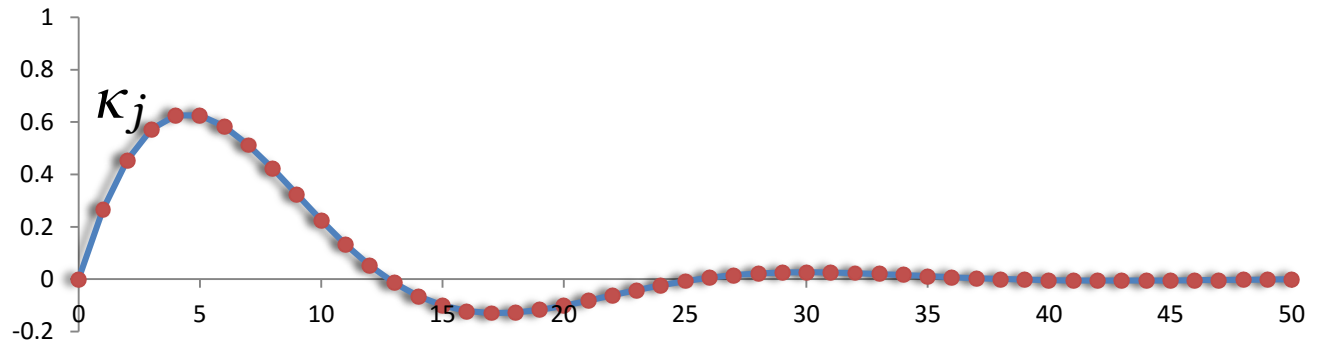
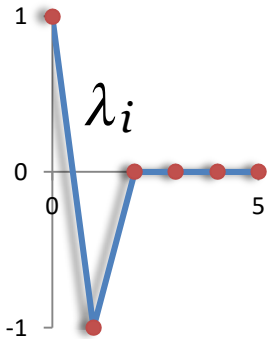
# Experiment settings

Wiener-Hammerstein system:



- Nonlinearity:  $m(w) = w^2$ .

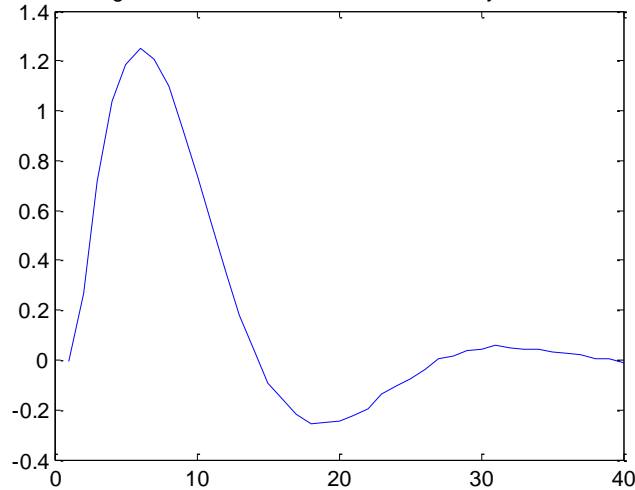
## Impulse responses:



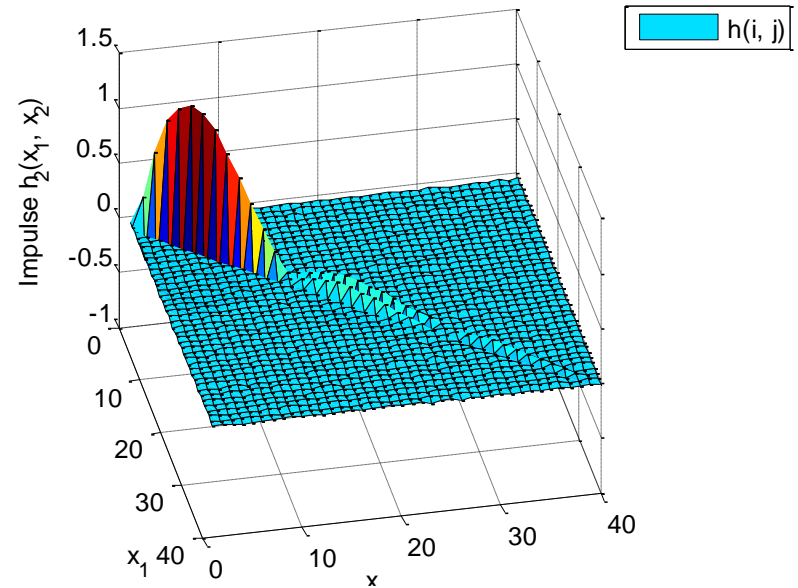
- Input signal: Uniform  $(-\sqrt{3}, \sqrt{3})$
- Noise signal: SNR=10
- Dictionary: Unique Volterra series components (L =40, P=2)

# Experiments results

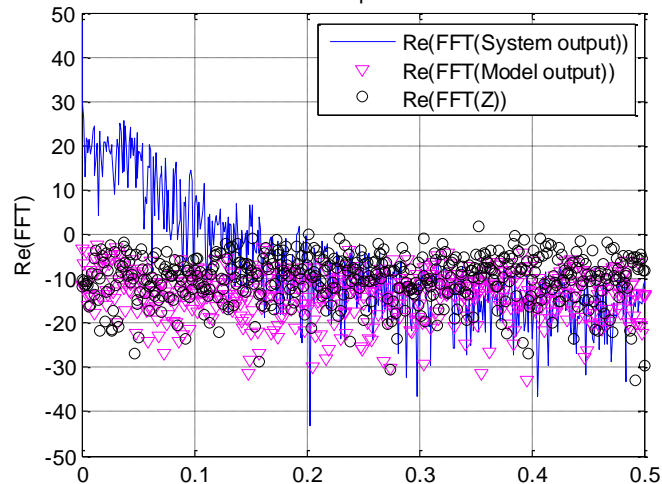
Diagonal of the 2nd order Volterra kernel by AGGREST



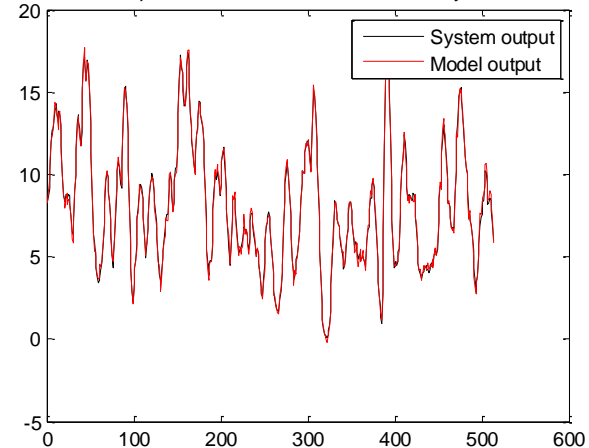
Second order Volterra kernel by AGGREST



FFT of measured validation output for N = 1024 and SNR = 100



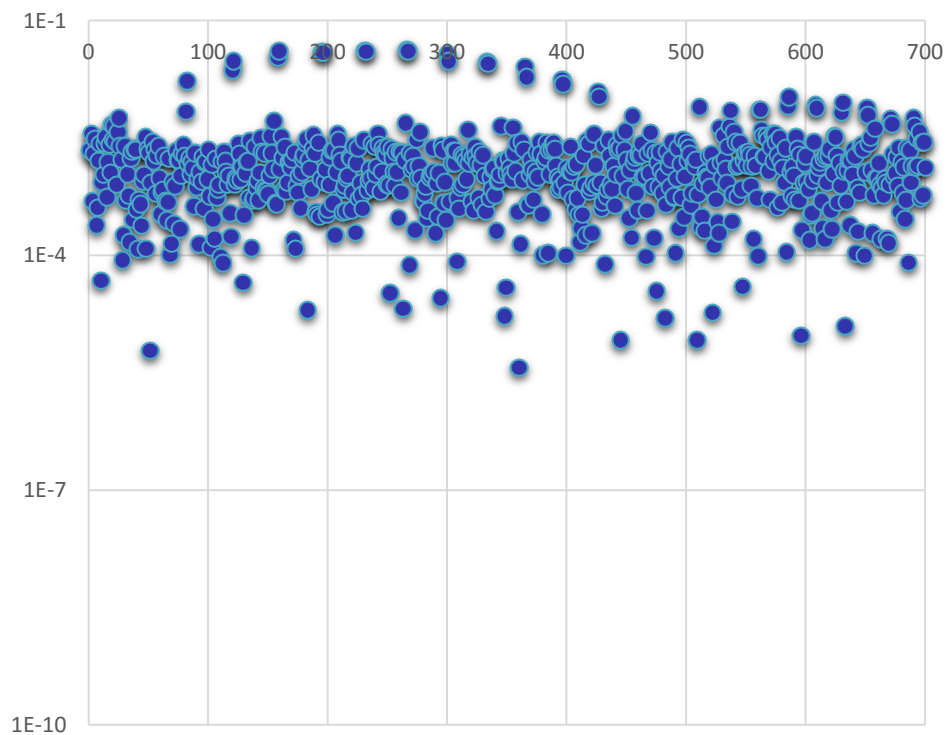
Outputs of AGGREST model and test system



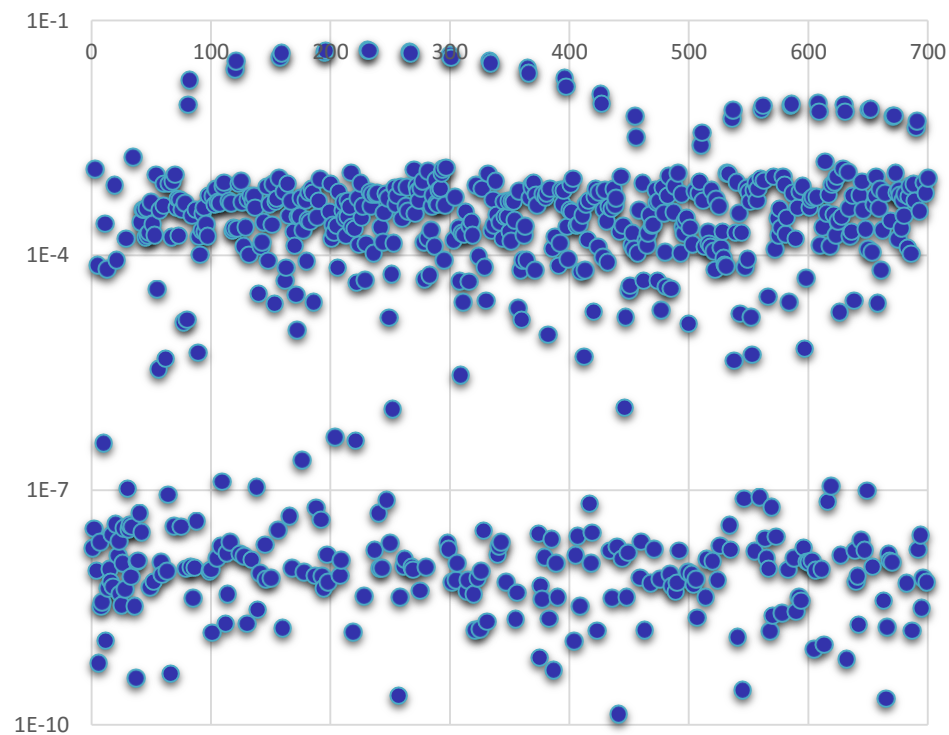
# LS Method vs Aggregative modeling

Dictionary weights evaluated by the LS method and aggregative approach.

LS METHOD



AGGREGATIVE APPROACH





# Bibliography

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Thank you